



Multi Area Load Frequency Control in Power Systems via Internal Model Control Scheme using Model-Order Reduction

J.Devendra Kumar¹, Mr.M.Santosh Kiran²

M.E Student [CS] , Dept. of EEE, ANITS College of Engineering , Visakhapatnam, A.P, India¹

Assistant professor, Dept. of EEE, ANITS College of Engineering, Visakhapatnam, A.P, India²

ABSTRACT: In an interconnected power system, as a power load demand varies randomly , both area frequency and tie-line power interchange also vary. When dealing with the LFC problem of power systems, unexpected external disturbances, parameter uncertainties and the model uncertainties of the power system pose big challenges for controller design. Present approach is based on two-degree-of-freedom, internal model control (IMC) scheme, which unifies the concept of model-order reduction like Routh and Padé approximations, and modified IMC filter design, recently developed by Liu and Gao [4]. The beauty of this paper is that in place of taking the full-order system for internal-model of IMC, a lower-order, i.e., second-order reduced system model, has been considered. This scheme achieves improved closed-loop system performance to counteract load disturbances. The proposed approach is simulated in MATLAB environment for a single-area power system consisting of single generating unit with a non-reheated turbine and due to its appreciable performance in a single area power system we have employed it to two area and four area power system. The effectiveness of the proposed controller is validated by applying a wide range of load disturbance.

Keywords: Internal model control (IMC), load frequency control (LFC), model-order reduction (MOR), robustness.

NOMENCLATURE OF POWER SYSTEM PARAMETERS

| | |
|-----------------|--|
| ΔP_d | Load disturbance (p.u.MW). |
| K_p | Electric system gain. |
| T_p | Electric system time constant (s). |
| T_T | Turbine time constant (s). |
| T_G | Governor time constant (s). |
| R | Speed regulation due to governor action (Hz/p.u.MW). |
| $\Delta f(t)$ | Incremental frequency deviation (Hz). |
| $\Delta P_G(t)$ | Incremental change generator in output (p.u MW) |
| ΔX_G | Incremental change in governor valve position. |

I.INTRODUCTION

Imbalances between load and generation must be within seconds to avoid frequency deviations that might threaten the stability and security of the power system. The problem of controlling the frequency in large power systems by adjusting the production of generating units in response to changes in the load is called load frequency control (LFC). The Objectives of LFC are to provide zero steady-state errors of frequency and tie-line exchange variations, high damping of frequency oscillations and decreasing overshoot of the disturbance so that the system is not to far from the stability. Many control strategies like Fuzzy logic PI and PID controllers[7] &[9], optimal control[3], Variable structure control[5], adaptive and self-tuning control[6] ,discrete time sliding mode control [8] , , and robust control [10] , [11] etc have been reported in the literature as an existing LFC solution. It is observed in power systems that the parameter values in the various power generating units like governors, turbines, generators, etc., fluctuate depending on system and power flow conditions which change almost every minute. Therefore, parameter uncertainty is an important issue for the choice of control technique. Hence, a robust strategy for LFC is required which takes care of both the uncertainties in system parameters and disturbance rejection.



International Journal of Advanced Research in Electrical, Electronics and Instrumentation Engineering

(An ISO 3297: 2007 Certified Organization)

Vol. 2, Issue 12, December 2013

In this paper the same method of IMC controller design via model order reduction(MOR) as in [2] is adapted to two-area and four area power system having identical areas with non-reheat turbines and compared with different controller configurations and the system is subjected to different load changes i.e. different cases of load changes in single area, and a case of load change in different areas. Initially the system frequency deviations are zero before any disturbance, a step load change is applied to a single area in first case and system behavior is observed without any controller, after then the proposed controller with only frequency change as input is used and the results are observed in comparison with previous ones, In all the cases it is observed that the negative overshoot is reduced monotonically, and damping of the oscillations is also increased. Hence this acceptable performance in single area power system gave an idea of extending this work to multi i.e. two and four area power system and examining its performance in multi area .

II. MOTIVATION AND PROBLEM STATEMENT

In order to design a robust controller for LFC problem, various control strategies as mentioned in the introduction section are useful. However, one class of strongly directional control strategy that has received extensive research in electric power components and process engineering is internal model control (IMC) [12]–[15]. This class of control technique is known to exhibit robustness, sub-optimality, less computational burden, and analytical as well as easily understandable approach. However, IMC has a little edge in comparison to aforementioned techniques with reference to command following and disturbance rejection. In literature, it is reported that it is also possible to optimise system performance for load disturbance rejection without sacrificing nominal set-point tracking using two degree- of-freedom (TDF) IMC [16]–[17].

As far as power system is concerned , one issue is that the inter connection of power systems result in huge increase in both the order of the system and number of controllers. With the ever-growing complexity of power systems in the electricity generation industry, formation of reduced-order models of these large-scale systems are extremely important. So in such cases, model-order reduction plays an important role in simplifying the design and implementation of the control systems. Moreover, as the size of model reduces, its computational complexity, size, and cost reduces. In [2], Saxena and Yogesh has proposed new strategy of IMC design via model order reduction(MOR) for single-area power systems, can also fulfil the control objectives in a satisfactory manner. So, this work motivated us to evaluate this IMC based controller using model-order reduction scheme for internal-model of a plant and extending it to multi area power system. Therefore, this new control strategy for LFC which is proposed which is a combination of modified IMC filter design and model-order reduction designed controller is capable of handling plant/model mismatches and parameter uncertainties.

More specifically, we aim to accomplish the following research objectives:-

a)Reduce the order of single-area power system having non- reheated type turbine. For simplicity, model-order reduction scheme of Padé [18] and Routh approximations [19] are applied. These reduced order models are treated as internal (predictive) models for imc structure

b)Consider TDF-IMC structure to optimize the performance of system for load disturbance rejection. The structure and proposed controller synthesis scheme is employed to demonstrate the effectiveness of utilizing reduced order models.

c)Evaluating it for single-area power system and extending it for two and four area power system.

International Journal of Advanced Research in Electrical, Electronics and Instrumentation Engineering

(An ISO 3297: 2007 Certified Organization)

Vol. 2, Issue 12, December 2013

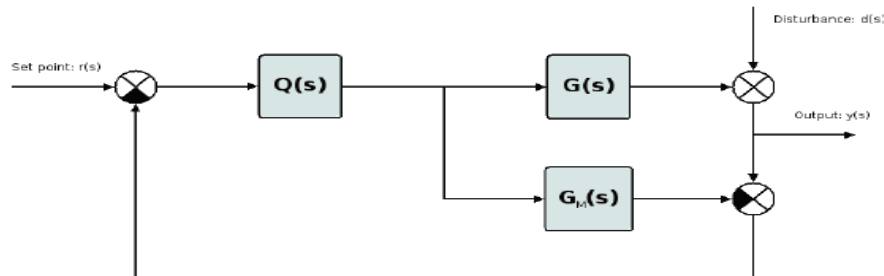


Fig. 1. Basic IMC structure.

III. IMC THEORY AND MODEL ORDER REDUCTION

The schematic representation of IMC structure is presented in Fig.1. The structure is characterized by a control device consisting of the feedback controller $Q(s)$, the real plant to be controlled $G(s)$, and a predictive model of the plant, i.e., the internal-model $G_M(s)$. The internal-model loop uses the difference between the outputs $G(s)$ of $G_M(s)$. This difference commonly known as an error, represents the effect of disturbances $D(s)$ and plant/model mismatch if exists. The two-step procedure for designing IMC controller is

1) Factor the model as

$$G_M(s) = G_{M+}(s)G_{M-}(s) \quad (1)$$

such that $G_{M+}(s)$ is a non-minimum phase part and $G_{M-}(s)$ is a minimum phase.

2) Define the IMC controller as

$$Q(s) = G_{M-}^{-1}(s)F(s) \quad (2)$$

Where $F(s)$ is a low-pass filter, commonly of the form

$$F(s) = (1 + \lambda s)^{-n} \quad (3)$$

In(3), λ is a tuning parameter, which adjusts the speed of response of a closed-loop system, and also removes plant/model mismatch which generally occurs at high frequency, thus responsible for robustness, n is an integer, chosen such that $Q(s)$ becomes proper/semi-proper for physical realization.

Two-Degree-of-Freedom IMC Controller

IMC scheme is based on pole-zero cancellation. It can achieve very good tracking ability; however, the response to disturbance rejection may be sluggish. So, a trade-off is required, where the performance for load disturbance rejection occurs by sacrificing set-point tracking. To avoid this problem, two different controllers $Q_D(s)$ and $Q_1(s)$, as shown in Fig. 2, are introduced in basic IMC structure [4]. Now, the set-point response and disturbance response of the modified IMC structure namely TDF-IMC, can be improved, and each In this presented work, we have considered the TDF-IMC structure as shown in Fig. 2, and applied the design scheme recently developed by Liu and Gao [4]. In Fig. 2, we can define $Q_D(s)$ as a disturbance rejection filter (feedback controller) and $Q_1(s)$ as a set-point filter. The closed-loop complementary sensitivity function $T(s)$ and multiplicative error $\mathcal{E}(s)$ which is a measure of plant/model mismatch can be defined respectively by

$$T(s) = Q_D(s)G_M(s) \quad (4)$$

and

$$\mathcal{E}(s) = \frac{(G(s)-G_M(s))}{G_M(s)} \quad (5)$$

International Journal of Advanced Research in Electrical, Electronics and Instrumentation Engineering

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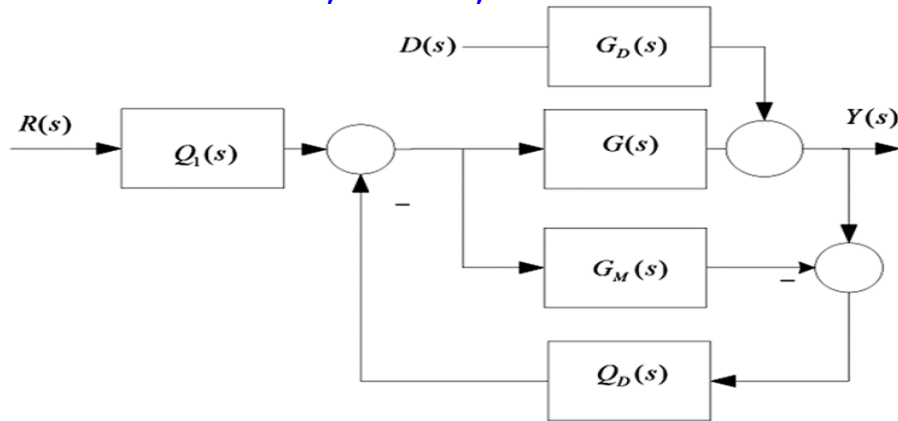


Fig. 2. TDF-IMC structure.

Since an effective IMC filter suggested in [24] is adopted to design IMC based controller for second-order internal-model of a system, therefore $F(s)$, of the form (3) is replaced by a modified filter $F'(s)$ such that

$$F'(s) = \frac{(\Psi s^2 + \theta s + 1)}{(\lambda_f s + 1)^x} \quad (6)$$

Where $x = 3$ or 4 , depending upon the requirement to make controller proper. On substituting (6) into (2), the TDF-IMC controller can be derived as

$$Q_D(s) = \frac{G_M^{-1}(\Psi s^2 + \theta s + 1)}{(\lambda_f s + 1)^x} \quad (7)$$

Where Ψ, θ should satisfy the following condition for each pole, p_1 and p_2 of the second-order system:

$$\lim_{s \rightarrow -p_i} (1 - T(s)) = 0 \quad (8)$$

$$T(s) = \frac{G_{M+}(s)(\Psi s^2 + \theta s + 1)}{(\lambda_f s + 1)^x} \quad (9)$$

Now, from (9), three cases arises for $G_{M+}(s)$:

1) *Case I:* When $G_{M+}(s)$ contains delay term only, i.e $G_{M+}(s) = e^{-\sigma s}$, then put $x=4$, and by substituting (9) into (8), we get

$$\Psi = \frac{p_1 e^{-\sigma p_2} (p_2 \lambda_f - 1)^4 - p_2 e^{-\sigma p_1} (p_1 \lambda_f - 1)^4 - p_1 + p_2}{p_1 p_2 (p_2 - p_1)} \quad (10)$$

$$\theta = \frac{p_1^2 e^{-\sigma p_2} (p_2 \lambda_f - 1)^4 - p_2^2 e^{-\sigma p_1} (p_1 \lambda_f - 1)^4 - p_1^2 + p_2^2}{p_1 p_2 (p_2 - p_1)} \quad (11)$$

2) *Case II:* When $G_{M+}(s)$ contains non-minimum phase term, then factorize $G_M(s)$ such that has $G_{M+}(s)$ only all-pass term, i.e., $G_{M+}(s) = \frac{1-as}{1+as}$ then put $x=3$, and by substituting (9) into (8), we get (12) and (13) as

$$\Psi = \frac{a^2 \lambda_f p_1 p_2 (p_1 + p_2) + (a \lambda_f^3 + 3a^2 \lambda_f^2) p_1 p_2 + a \lambda_f (p_1^2 + p_2^2 + p_1 p_2) + (\lambda_f^3 + 3a \lambda_f^2) (p_1 + p_2) + 3 \lambda_f^2}{a^2 p_1 p_2 + a (p_1 + p_2 + 1)} \quad (12)$$

$$\theta = \frac{a \lambda_f^2 p_1 p_2 + (3a \lambda_f^2 + 3a^2 \lambda_f) p_1 p_2 - 3a \lambda_f (p_1 + p_2) + a \lambda_f p_1 p_2 (p_1 + p_2) + (\lambda_f^2 + 3a \lambda_f) p_1 p_2 - 3 \lambda_f}{a^2 p_1 p_2 + a (p_1 + p_2 + 1)} \quad (13)$$

3) *Case III:* When $G_{M+}(s)$ neither contains non-minimum phase term nor delay term, i.e., $G_{M+}(s) = 1$ then it can be considered as a special case of above mentioned *case I*. Therefore, on substituting $\sigma=0$, in (10) and (11), brings

$$\Psi = \frac{p_1 (p_2 \lambda_f - 1)^4 - p_2 (p_1 \lambda_f - 1)^4 - p_1 + p_2}{p_1 p_2 (p_2 - p_1)} \quad (14)$$

$$\theta = \frac{p_1^2 (p_2 \lambda_f - 1)^4 - p_2^2 (p_1 \lambda_f - 1)^4 - p_1^2 + p_2^2}{p_1 p_2 (p_2 - p_1)} \quad (15)$$



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Thus, it is clear that controller $Q_D(s)$ expressed by (7) does not require heavy computational burden. Hence, controller's simplicity and easy practical implementation are the major advantage of this design scheme. Here, we are only concerned with disturbance rejection problem, i.e., effect of $D(s)$ on $Y(s)$ we need not to evaluate set-point $Q_1(s)$ filter since $R(s) = 0$ is assumed.

B. Model Order Reduction

By model-order reduction we mean roughly that the large scale system (higher-order or full-order) is approximated by small-scale system (lower-order or reduced-order), such that the inherent behavior of the original system does not deteriorate. In terms of control perspective, the basic concept behind the model-order reduction technique is to preserve the dominant poles of the full-order model of the plant while rejecting the non-dominant poles. Until now, many model reduction techniques have been developed [18]–[24]. These methods can be utilized for SISO/MIMO systems to obtain lower-order models which further can be used to design IMC based controller. As an example of utilizing reduced order modeling, only two methods: Padé and Routh approximations [18], [19] are considered in the present study. The application of these techniques to LFC is elaborated in Section V.

IV. PROPOSED IMC STRATEGY

As discussed earlier in order to apply IMC design scheme a perfect model is required. Furthermore, the controller must be able to invert the model perfectly. However, in real time applications, it is difficult to get a perfect model. So, generally the process is approximated as first-order or second-order plus dead time (FOPDT or SOPDT) model. This results in addition of delay terms in the transfer function. Since the IMC controller needs inverse plant model, and the inversion of delay terms for controller design leads to predictor action. Moreover, most often the obtained transfer function is of higher-order which sometimes leads to unrealizable controller and results into slower response, and more complex computation. Thus, there is a need of model-order reduction techniques to develop causal, realizable, and lower-order process models.

So, based on Sections III, the proposed IMC design involves following two steps

- 1) Approximate the model of a system using Padé or Routh approximation techniques.
- 2) Evaluate the TDF-IMC controller $Q_D(s)$ for this approximated (reduced) model.

V. LFC FOR SINGLE-AREA POWER PLANT

A) Plant Description

Usually, the power systems are large-scale systems with complex nonlinear dynamics [25]. However, for relatively load disturbance, they can be linearized around the operating point. Here, a single-area power system supplying power to a single service-area through single generator is considered. This power plant for LFC design consists of governor $G_g(s)$, non-reheated turbine $G_t(s)$, load and machine $G_p(s)$, and $1/R$ is the droop characteristics, a kind of feedback gain to improve the damping properties of the power system. The linear model of plant is shown in Fig. 3. The dynamics of these subsystems are

$$\begin{aligned} G_g(s) &= \frac{1}{T_G s + 1} \\ G_t(s) &= \frac{1}{T_T s + 1} \\ G_p(s) &= \frac{K_p}{T_p s + 1} \end{aligned} \quad (16)$$

The whole system model can be illustrated by

$$\Delta f(s) = \mathbf{G}(s)\mathbf{u}(s) + \mathbf{G}_d(s)\Delta P_d(s) \quad (17)$$

$$\begin{aligned} G_d(s) &= \frac{G_g(s)G_t(s)G_p(s)}{1 + \frac{G_g(s)G_t(s)G_p(s)}{R}} \\ &= \frac{K_p}{T_T T_p T_G s^3 + (T_p T_T + T_T T_G + T_p T_G) s^2 + (T_p + T_T + T_G) s + (1 + \frac{K_p}{R})} \end{aligned} \quad (18)$$

$$G_d(s) = \frac{G_p(s)}{1 + G_g(s)G_t(s)G_p(s)/R} \quad (19)$$

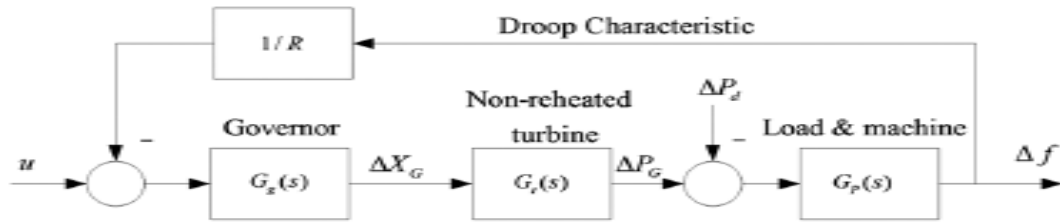


Fig.3. Linear model of a single- area power system

Equation (17) clearly explains that LFC is basically a disturbance rejection (regulator) problem in which the objective is to evaluate the control law $u(s) = -K(s)\Delta f(s)$ where $K(s)$ is IMC based compensator to control the power plant $G(s)$ and minimize effect on $\Delta f(s)$ in the environment of small load disturbance $\Delta P_D(s)$ [26].

B. Model-Order Reduction of Plant

It is clear from (18) that even the single-area power system containing only one generator, still, it is of third-order, and thus IMC control design is obviously of higher order if the full-order model is used. So, we obtain the second-order reduced-model of the single-area power system using following methods.

1) Padé Approximation Method : This reduction method is based on matching of few coefficients of Taylor series expansion, about $s=0$ of the reduced order model with the corresponding coefficients of the original model. In order to convert higher-order system $G(s)$ into second-order reduced model $G_{MR}^{pade}(s)$, we first define $G_{MR}^{pade}(s)$ as

$$G_{MR}^{pade}(s) = \frac{(a_0 + a_1s)}{(b_0 + b_1s + s^2)} \tag{20}$$

Equation (18) can be rewritten as

$$G(s) = \frac{K_p/A}{s^3 + (B/A)s^2 + (C/A)s + (D/A)} \tag{21}$$

where

$$\begin{aligned} A &= T_p T_T T_G, \quad B = T_p T_T + T_T T_G + T_p T_G \\ C &= T_p + T_T + T_G, \quad D = 1 + \frac{K_p}{R} \end{aligned} \tag{22}$$

The coefficients of the power series expansion $G(s)$ can be expressed as $G(s) = C_0 + C_1s + C_2s^2 + C_3s^3 + \dots$, which yields

$$c_0 = \frac{K_p}{D}, \quad c_1 = \frac{-CK_p}{D^2}, \quad c_2 = \frac{(C^2 - BD)K_p}{D^3}, \quad c_3 = \frac{(2BCD - AD^2 - C^3)K_p}{D^4} \tag{23}$$

Now, to obtain second-order reduced model, $G_{MR}^{pade}(s)$ of the form described in (20), the parameters a_i and $b_i (i = 0,1)$ can be evaluated by simplifying

$$\begin{pmatrix} c_2 & c_1 \\ c_3 & c_2 \end{pmatrix} \begin{pmatrix} b_0 \\ b_1 \end{pmatrix} = \begin{pmatrix} -c_0 \\ -c_1 \end{pmatrix}, \quad \begin{pmatrix} a_0 \\ a_1 \end{pmatrix} = \begin{pmatrix} c_0 & 0 \\ c_1 & c_0 \end{pmatrix} \begin{pmatrix} b_0 \\ b_1 \end{pmatrix} \tag{24}$$

2) Routh Approximation Method: In this method, the reduced order model can be obtained by approximating the coefficients of Routh table. Consider a second-order reduced-model $G_{MR}^{routh}(s)$ as $G_{MR}^{routh}(s) = \frac{P_2(s)}{Q_2(s)}$ where

$P_2(s)$ and $Q_2(s)$ are numerator and denominator, respectively. We first reciprocate $G_{MR}^{routh}(s)$ using relation $L\tilde{(s)} = \left(\frac{1}{s}\right)L\left(\frac{1}{s}\right)$. Thus, the reciprocated model of $G\tilde{(s)}$ becomes

$$G\tilde{(s)} = \frac{K_p s^2}{(Ds^3 + Cs^2 + Bs + A)} \tag{25}$$

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and then, expand $G^{\sim}(s)$ namely

$$G^{\sim}(s) = \frac{P_1^{\sim}(s)}{Q_1^{\sim}(s)} = \sum_{t=1}^n \beta_t \prod_{j=1}^t F_j(s) \quad (26)$$

Where $\beta_i (i = 1,2)$ are constants, $F_i(s) (i = 1,2)$ and contains α_i terms. Next, we need to compute α and β tables corresponding to $G^{\sim}(s)$, which is shown in Table I. The detailed study and evaluation of α and β tables are reported in [19]. These α and β terms gives reciprocated reduced-order numerator $P_2^{\sim}(s)$ and $Q_2^{\sim}(s)$ denominator for second-order reduced model as

TABLE I: α - β TABLE FOR ROUTH APPROXIMATION

| | α Table | | β Table | |
|----------------------------------|--------------------|-----|---------------------------|-----|
| $\alpha_1 = \frac{D}{C}$ | D | B | K_p | 0 |
| | C | A | $\beta_1 = \frac{K_p}{C}$ | 0 |
| $\alpha_2 = \frac{C^2}{BC - AD}$ | $B - \frac{AD}{C}$ | | $\beta_2 = 0$ | 0 |

$$\begin{aligned} P_2^{\sim}(s) &= \beta_2 + \alpha_2 \beta_1 s \\ Q_2^{\sim}(s) &= 1 + \alpha_2 s + \alpha_1 \alpha_2 s^2 \end{aligned} \quad (27)$$

On substituting values of α and β in (27), we get

$$P_2^{\sim}(s) = \frac{CK_p s}{(BC - AD)} \quad Q_2^{\sim}(s) = 1 + C^2 s / (BC - AD) + CD s^2 / (BC - AD) \quad (28)$$

Finally, the required reduced order model is obtained by again reciprocating the terms of (28), which gives

$$G_{MR}^{Routh}(s) = \frac{CK_p}{((BC - AD)s^2 + C^2 s + CD)} \quad (29)$$

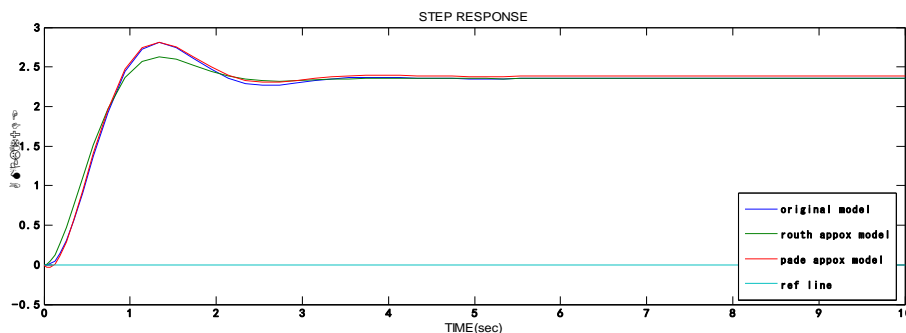


Fig. 4. Comparison between step responses of original model and that of reduced-order model

VI. SIMULATION STUDIES

Consider the typical values of parameters for single-area power system as expressed in [1]:

$$K_p = 120 ; T_T = 0.3 ; T_G = 0.08 ; T_P = 20 ; R = 2.4; \quad (30)$$

Using (30), $G(s)$ is evaluated as



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$$G(s) = \frac{250}{s^3 + 15.88s^2 + 42.46s + 106.2} \quad (31)$$

which is a third-order under-damped system. Since IMC requires a plant model in its control structure, so, before applying the proposed scheme as mentioned in Section IV, consider the predictive model $G_M(s)$ for IMC structure same as the original full-order power plant, i.e., $G_M(s) = G(s)$. The various steps to provide the proposed design scheme are as follows.

A. Application of Model-Order Reduction

Using Padé approximation and Routh approximation method, the second-order reduced models of (31) are

$$G_{MR}^{pade}(s) = \frac{-1.191s + 18.92}{s^2 + 2.708s + 8.043} \quad (32)$$

$$G_{MR}^{routh} = \frac{18.68}{s^2 + 3.173s + 7.94} \quad (33)$$

The step responses of the original model, i.e., full-order model $G(s)$ and reduced order models expressed in (32) and (33), respectively, are shown in Fig. 4. From this figure, it is evident that the response of the original third-order model is almost equal to that of reduced second-order models. Thus, we can say that two models are in good approximation.

B. Application of Proposed Controller Design

1) Controller for Padé Approximation Model: Since (32) has RHP zero at $s=15.89$, and therefore in order to factorize (32), $G_{MR}^{pade}(s)$ can be written as

$$G_{MR}^{pade}(s) = \frac{(1.191s + 18.92)}{(s^2 + 2.708s + 8.043)} \frac{(-1.191s + 18.92)}{(1.191s + 18.92)} \quad (34)$$

Where $G_{MR-}^{pade}(s)$ is a minimum phase part:

$$G_{MR-}^{pade}(s) = \frac{(1.191s + 18.92)}{s^2 + 2.708s + 8.043} \quad (35)$$

and $G_{MR+}^{pade}(s)$ is a non-minimum phase part:

$$G_{MR+}^{pade}(s) = \frac{(-1.191s + 18.92)}{(1.191s + 18.92)} \quad (36)$$

Taking $\lambda_f = 0.08$, and using (12) and (13), the TDF-IMC controller of the form (7) is given by
Where Ψ , θ , λ_f and x are 0.0057, 0.1687 and 3, respectively.

$$Q_D^{pade}(s) = \frac{(s^2 + 2.708s + 8.043)(0.0057s^2 + 0.1687s + 1)}{(1.191s + 18.92)(0.08s + 1)^3} \quad (37)$$

2) Controller for Routh Approximation Model: For evaluating TDF-IMC controller when Routh approximated reduced second-order model expressed in (33) is used, there is no need to factorize (33) because it does not contain any RHP zero or delay factor. So, in this case, $G_{MR}^{routh}(s) = G_{MR-}^{routh}(s)$ Now, $\lambda_f = 0.2$ selecting, and using (14) and (15), the controller is given by

$$Q_D^{routh}(s) = \frac{(s^2 + 3.173s + 7.94)(0.1419s^2 + 0.5862s + 1)}{(18.68)(0.2s + 1)^4} \quad (38)$$

where Ψ , θ and x are 0.1419, 0.5862, and 4, respectively.

Here we have applied a non-periodic load disturbance $\Delta P_D(t) = 0.01$ at $t = 2$ sec as shown in Fig. 3.

C. Performance Evaluation and Comparative Remarks

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The effectiveness of the resulting controller is compared with the IMC tuned PID controller developed by Wen Tan [1]. The disturbance rejection response of the power system for nominal case of imc tuned pid, Padé and Routh approximation models are illustrated in Fig. 6. The comparison of the three response reveals that Routh and Padé approximated model are efficient models to obtain TDF-IMC controllers and reaches frequency deviation zero faster.

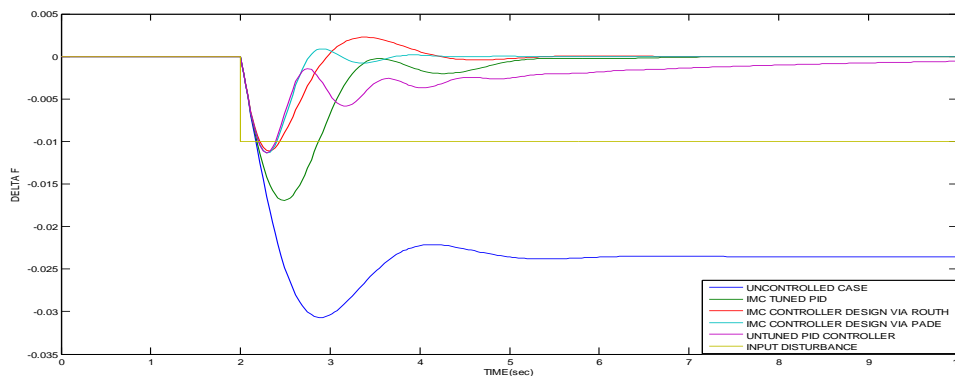


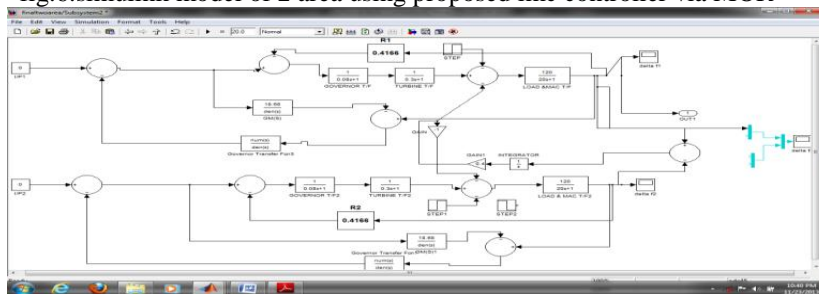
Fig.5.responses of power system for different control techniques

VII. SIMULATED MODEL OF TWO AND FOUR AREA POWER SYSTEM

A.TWO AREA POWER SYSTEM:

A two area simulink model using proposed controller which is adapted in the work is shown in fig.6. Each area is assumed to have only one equivalent generator and is equipped with governor- turbine system. The terms showed in the figure are termed in the nomenclature above.

fig.6.simulink model of 2 area using proposed imc controller via MOR



The output responses of two area power system with non reheat turbine for varies configuration is shown below figure

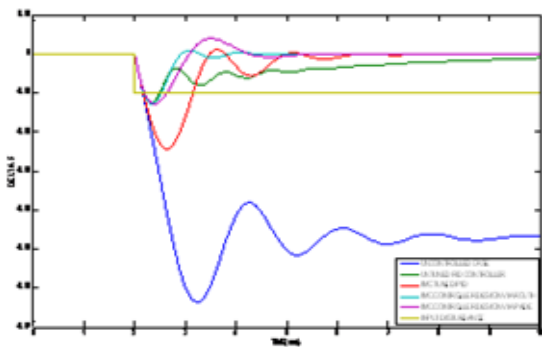


Fig.7. Δf variation in area 1 due to load change $\Delta Pd1=0.01$ and $\Delta Pd2=0.03$ in both areas respectively

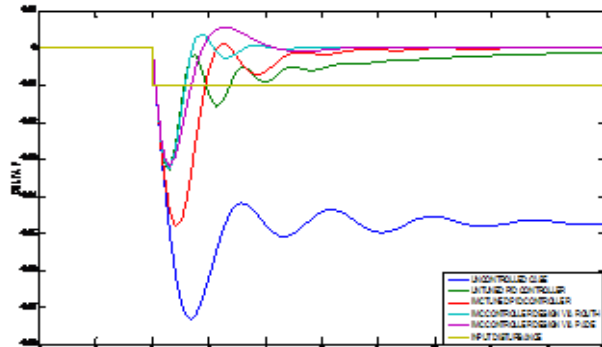


Fig.8. Δf variation in area 2 due to load change $\Delta Pd1=0.01$ and $\Delta Pd2=0.03$ in both areas

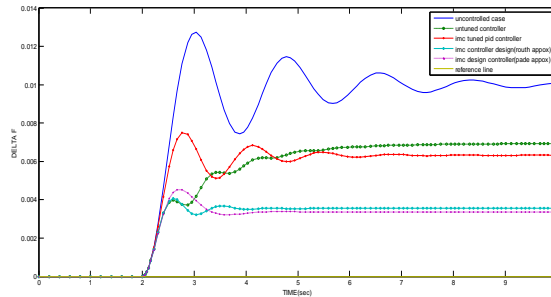


Fig.9. ΔP_{tie} variation due to load change $\Delta P_{d1}=0.01$ and $\Delta P_{d2}=0.03$ in both areas

B.FOUR-AREA POWER SYSTEM

Power systems have variable and complicated characteristics and comprise different control parts and also many of the parts are nonlinear [1]. These parts are connected to each other by tie lines and need controllability of frequency and power flow. Interconnected multiple-area power systems can be depicted by using circles. A simplified four area interconnected power system used in this study is shown in Fig. 1.

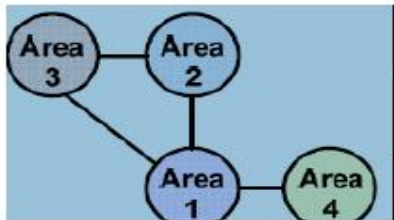
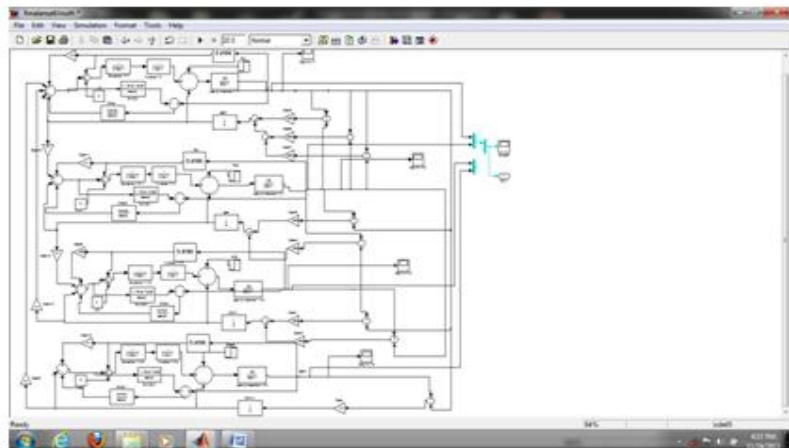


Fig. 10. Simplified interconnected power system



In Fig.11, a four-area interconnected system simulink model is depicted

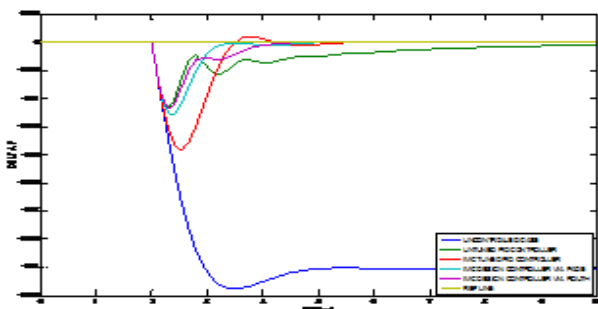


Fig.12. Δf variation in area1 due to load change $\Delta P_{di}=0.01$ in all four areas

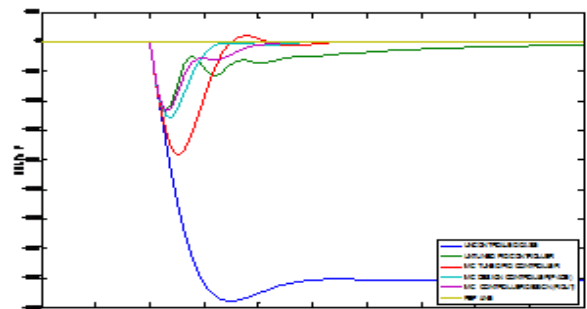


Fig.13. Δf variation in area2 due to load change $\Delta P_{di}=0.02$ in all four areas

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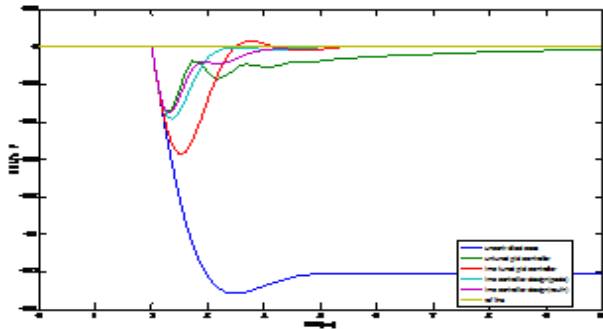


Fig.14. Δf variation in area3 due to load change $\Delta P_{di}=0.03$ in all four areas

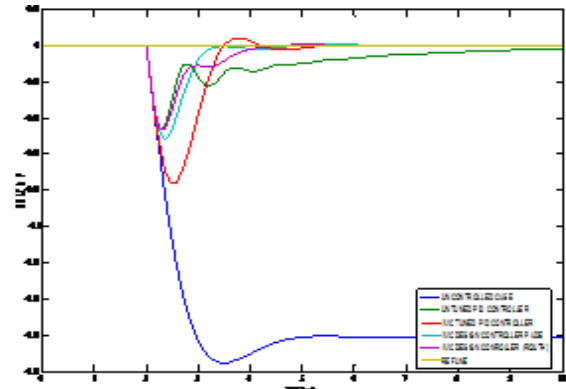


Fig.15. Δf variation in area4 due to load change $\Delta P_{di}=0.04$ in all four areas

VIII. ROBUSTNESS AGAINST UNCERTAINTY IN MODEL PARAMETERS

The disturbance response of the power system without any control for nominal as well as uncertain models are shown in Fig.16, which states that the disturbance at output is approximately 50% higher for nominal case, and 100%, 150% more for +50% and -50% uncertain model, respectively, as compared to the input disturbance. In such cases, it is essential to confirm whether the same controller can handle all such parameter uncertainties. Here, we have considered 50% additive uncertainty in all the parameters.

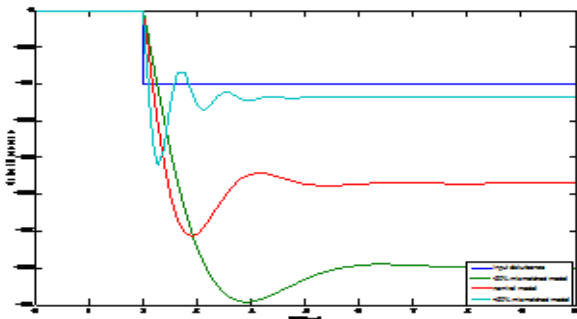


Fig.16. Effect of disturbance at output for nominal and uncertain models.

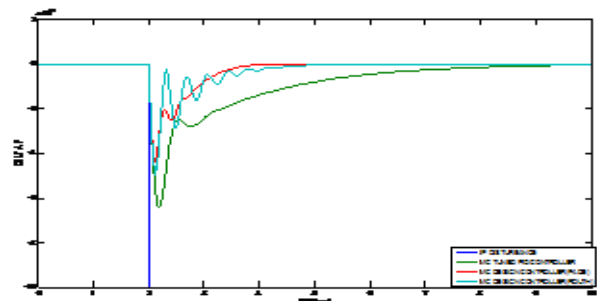


Fig.17. Responses of a power system using TDF-IMC design with various reduced-order models for (a) lower bound and (b) upper bound uncertainties

Fig. 17(a) and (b) shows the disturbance rejection response for +50% and -50% uncertain systems. Thus, it is evident that the same controllers, expressed in (37)–(38), for each model are indeed capable of handling parameter variations, and achieve superior performance compared to the controller proposed in [23]. Thus, the proposed schemes are robust in nature.

VII. CONCLUSION

In electricity power industry, there is an ongoing need for efficient and effective LFC techniques to counter the ever-increasing complexity of large-scale power systems. As seen from the results, for the uncontrolled case which is represented in solid line has more negative overshoot and more oscillations, which are decreased by applying different proposed controllers and configurations. A TDF-IMC controller via MOR (using routh and pade approximation) is used for load frequency controller of single area power system. The comparative study shows that system performance characteristics of TDF-IMC via MOR method (proposed controllers) is better than IMC TUNED PID controller or untuned pid controller and is more effective in



International Journal of Advanced Research in Electrical, Electronics and Instrumentation Engineering

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reducing the frequency deviation transients, maintain the robust performance, minimize the effect of disturbances and specified uncertainties, very effectively.

FUTURE SCOPE

Hence future scope of present work is that to investigate the efficient model order reduction techniques for achieving better approximation to full order system and effective performances of the system.

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